

# The Component Placement Process in Surface Mount Technology Based on Coordinate Calculation Equations

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## ABSTRACT

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Accurate component placement in Surface Mount Technology (SMT) is essential for assembly yield and long-term product reliability. In practice, placement accuracy is often degraded by fixture offsets, vibration, and feeder/nozzle variability, while many optimization approaches remain predominantly offline and therefore cannot directly compensate for runtime deviations. This paper proposes a generalized coordinate calculation framework based on linear transformation principles, implemented in a PLC-controlled three-axis pick-and-place prototype. Using a small set of reference parameters, the proposed method generates consistent coordinates for passive components, ICs, and feeders and supports fast recalculation when the PCB reference origin changes. Experimental validation was conducted on 62 PCB assemblies, achieving a placement success rate of 91.94% (57/62). For all successful boards, the placement deviation remained within the predefined tolerance window ( $\pm 0.2$  mm), and observed failures were attributable to hardware issues (feeder misfeed and vacuum loss) rather than the coordinate model. The research contribution of this paper is a practical, real-time adaptable alternative for coordinate programming in SMT pick-and-place systems, with future work focusing on vision-assisted referencing, hardware reliability improvements, and benchmarking against offline optimization methods on larger PCB layouts.

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## 1. Introduction

In modern electronics manufacturing, automated component placement machines employing Surface Mount Technology (SMT) are widely utilized [1]–[3], playing a pivotal role in ensuring placement accuracy and assembly efficiency compared to conventional through-hole technology [4]–[6]. Nevertheless, one of the most critical challenges in SMT systems lies in achieving highly precise component coordinate calculation with minimal error [7], particularly for ultra-miniaturized components. Accurate coordinate determination not only influences soldering quality but also directly impacts the reliability and lifespan of the final product [8], [9].

In the current context, where the integration density of printed circuit boards (PCBs) continues to increase [10]–[13], the precision requirements for the pick-and-place process have advanced beyond the micrometer scale, approaching the nanometer level for certain high-end electronic devices [14]. This trend makes coordinate determination and trajectory optimization of the placement nozzle a crucial issue for manufacturers. Furthermore, the rapid growth of the Internet of Things (IoT), mobile devices, and embedded systems has created a strong demand for compact PCBs densely populated with complex components, thereby introducing additional challenges to the assembly process [15]–[19].

The assembly of a complete printed circuit board (PCB) requires multiple stages, including adhesive application, component pick-and-place, reflow soldering, and inspection [20], [21]. Among these, the SMT pick-and-place process is the most critical stage, as it demands near-perfect accuracy in positioning each component on the PCB [22]. Therefore, determining component coordinates using mathematical equations is essential for improving both placement accuracy and production reliability. In present, various approaches have been introduced to determine placement paths or component coordinates, such as Heuristic algorithms [23]–[26], Genetic algorithms [27]–[31], and Ant Colony algorithms [32]–[36], all aiming to enhance the accuracy of electronic component assembly.

However, evolutionary and metaheuristic algorithms primarily optimize parameters in an offline manner, meaning coordinate optimization must be completed prior to actual operation. This constitutes their major limitation, since in practical manufacturing environments, component positions may shift due to machine vibrations, fixture tolerances, or mechanical interactions, which can reduce placement accuracy and render offline optimization ineffective [37]–[41]. Hence, there remains a clear research gap in developing generalized coordinate calculation solutions that can dynamically adapt to positional variations under real operating conditions.

In addition to optimizing placement efficiency through feeder allocation and nozzle selection, the optimization of the assembly process by means of coordinate calculation is equally important for improving both performance and quality [25], [42]–[46]. To address these gaps, the present research develops a generalized mathematical model to accurately determine component coordinates and optimize placement paths in SMT systems, applied to a 3.3V low-voltage circuit Fig. 1. The proposed method reduces coordinate errors and shortens assembly time by enabling systematic coordinate generation and flexible adjustment of PCB positions according to actual machine conditions. As a result, this approach overcomes key limitations of previous evolutionary algorithms and provides a practical foundation for improving SMT placement performance in real production environments.

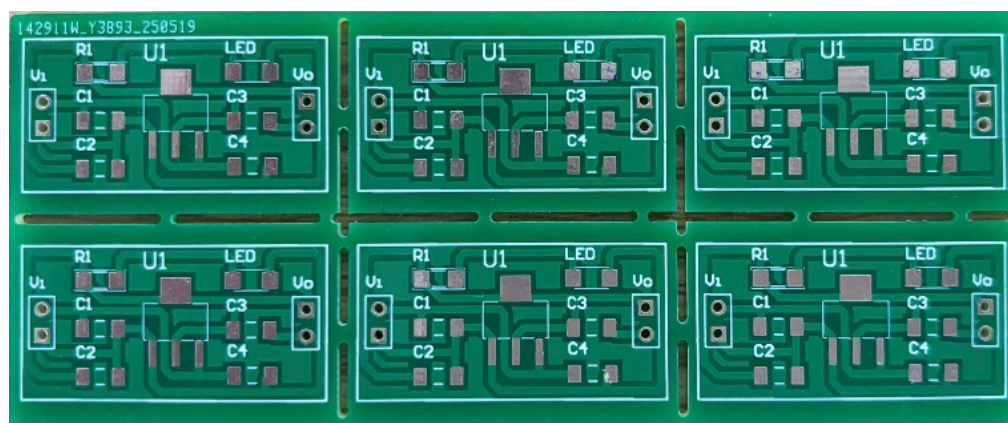


Fig. 1. Printed circuit board (PCB) used for processing

To address these gaps, the present research develops a generalized mathematical model to accurately determine component coordinates and optimize placement paths in SMT systems, applied to a 3.3V low-voltage circuit Fig. 1. This approach reduces errors and shortens assembly time. Furthermore, the proposed equations allow operators to flexibly adjust PCB positions according to the actual machine conditions, overcoming the limitations of previous evolutionary algorithms.

## 2. Component Coordinate Calculation on the PCB

The proposed method aims to reduce manual coordinate entry by generating all placement coordinates from a limited set of reference parameters. At runtime, the operator (or the system) provides a PCB reference point in machine coordinates (datum) and a small set of spacing parameters derived from the PCB layout. The algorithm then calculates coordinates for all placement points and feeder pickup locations.

Conceptually, the mapping from PCB reference coordinates to the machine coordinate system can be interpreted as a linear transformation dominated by translation and (if needed) a small planar rotation. In our PLC implementation, the mapping is realized using straight-line relationships of the form  $y = ax + b$  [47], which provides an efficient arithmetic form for coordinate generation and fast recalculation when the datum changes.

Fig. 2 illustrates the XY motion axes (horizontal plane), driven by two 200 W servo motors, each directly coupled to a 600 mm travel ball screw, a design that ensures high positioning accuracy [48]–[52]. In addition, the Z axis (vertical axis) controls the pick-and-place operations, driven by a 100 W servo motor combined with a belt-drive mechanism. This configuration allows the synchronized movement of two vacuum nozzles in a fast and flexible manner. The coordinate system employed in this study is depicted in Fig. 3, it provides a visual representation of the two key coordinate systems utilized throughout this study: the machine coordinate system, which defines the absolute working envelope of the equipment, and the PCB reference system, which is used to locate components relative to the board itself.

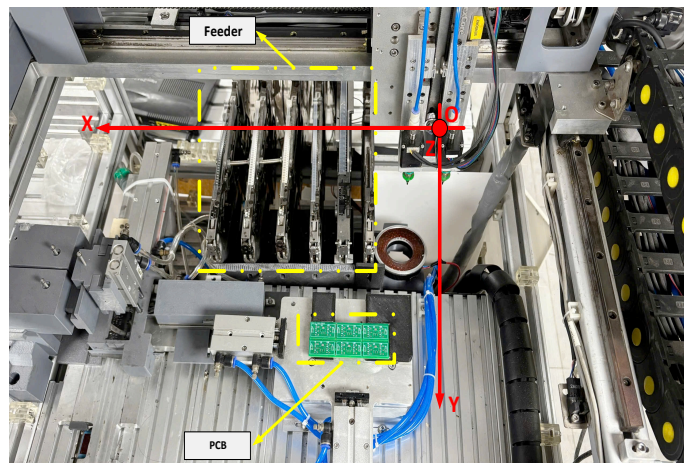


Fig. 2. Experimental SMT model

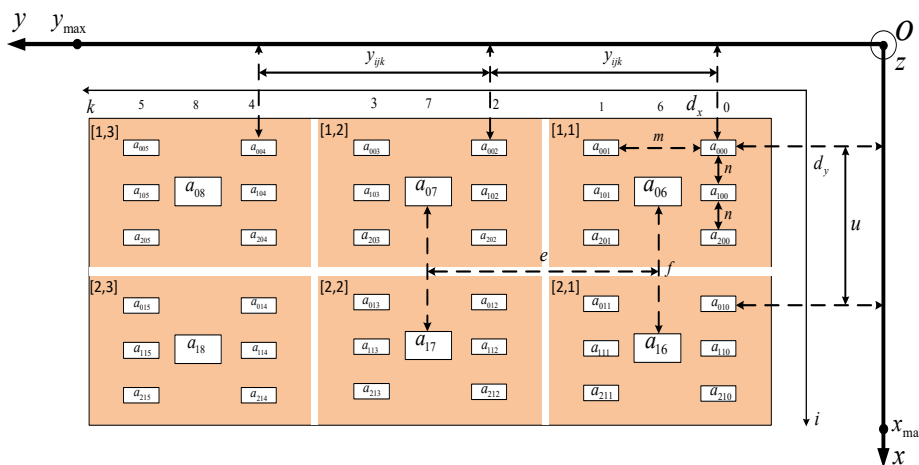


Fig. 3. Diagram of the machine coordinate system relative to the PCB and component position

where  $i$  is the iteration index along the  $x$ -axis;  $j$  is the number of components with respect to  $i$ ;  $k$  is the iteration index of components in column order along the  $y$ -axis;  $d_x$  and  $d_y$  is the current coordinate of the component to be placed at the first position on the PCB ( $a_{000}$ );  $x_{max}$  and  $y_{max}$  denotes the travel limit of the ball screw;  $y_{ij}$  is the spacing between components in successive PCBs (between the  $2k - th$  and  $(2k + 1) - th$  boards);  $x_j$  is the spacing between two consecutive components with respect to  $i$ ;  $n$  is the spacing between two components along the  $x$ -axis on a single PCB;  $m$  is the spacing between two components along the  $y$ -axis between two PCBs;  $e$  is the spacing between two ICs along the  $y$ -axis;  $f$  is the spacing between two ICs along the  $x$ -axis; and  $u$  is the spacing between two components along the  $x$ -axis between two PCBs.

## 2.1. Component Coordinates

To ensure the generality and reproducibility of the results, the current prototype is constructed based on the following assumptions:

- 1) **Rigid PCB assumption:** The printed circuit board (PCB) is considered perfectly rigid during pick-and-place operations. Therefore, the influence of board warpage and any warpage compensation mechanism is not included in the present study.
- 2) **Negligible scale effect:** Scale effects are neglected (i.e., the scale factor is approximately 1). Hence, the dominant deviations in practical operation are expected to originate mainly from translational displacement and rotational errors, which may be introduced by fixture inaccuracy and reference misalignment.
- 3) **Placement accuracy criterion:** The placement accuracy is evaluated using a predefined tolerance window of  $\pm 0.2$  mm. In the context of PCB footprint/land-pattern tolerance analysis, the IPC-7351 standard defines the placement tolerance parameter  $P$ , which is widely adopted as a reference in engineering practice.
- 4) **Unified linear representation:** To improve both generality and computational efficiency, the coordinates of components handled by the primary nozzle and the secondary nozzle are expressed using a general linear equation.

Based on this framework, the study plans the overall coordinates of all objects, such as discrete components, ICs, and feeders, on the Oxy plane. These are then converted into the form of coordinates represented by the general linear equation to facilitate programming, operation, and maintenance in subsequent implementation stages.

### 2.1.1. Linear Transformation Model for Component Mapping

The coordinate mapping is derived from the straight-line equation, where the slope coefficient  $a$  is computed from two reference points:

$$a = \frac{y_2 - y_1}{x_2 - x_1} = \frac{y_{max} - d_y}{x_{max} - d_x} \quad (1)$$

Here,  $x_1 = d_x$  and  $y_1 = d_y$  are the current coordinates of the component to be placed at the first position of the PCB;  $x_2 = x_{max}$  and  $y_2 = y_{max}$  represent the travel limits of the ball screw axes;  $y_{ij}$  denotes the spacing of components between the  $2k - th$  and  $(2k + 1)$ -th boards, while  $x_j$  refers to the spacing between two consecutive components with respect to index  $i$ , defined as:

$$x_j = \begin{cases} 0, & \text{when } j = 0 \\ u, & \text{when } j > 0 \end{cases}$$

Here,  $u$  denotes the spacing between two components along the  $x$ -axis across two printed circuit boards. The definitions are as follows:  $i$  represents the number of iterations along the  $x$ -axis, and  $j$  is the component index with respect to  $i$ .

### 2.1.2. Slope Coefficients for the First Column ( $k = 0$ )

To establish the coordinate coefficients for component placement, the slope coefficient  $a$  is first derived for the first PCB column ( $k = 0$ ) at successive component indices  $i = 0, 1, 2$ .

For the first component placement at the first column ( $i = 0, k = 0$ ):

$$\begin{cases} a_{00} = \frac{y_{\max} - d_y}{x_{\max} - x_0 - 0n - d_x} \\ a_{01} = \frac{y_{\max} - d_y}{x_{\max} - x_1 - 0n - d_x} \end{cases} \quad (2)$$

For the second component placement at the first column ( $i = 1, k = 0$ ):

$$\begin{cases} a_{10} = \frac{y_{\max} - d_y}{x_{\max} - x_0 - 1n - d_x} \\ a_{11} = \frac{y_{\max} - d_y}{x_{\max} - x_1 - 1n - d_x} \end{cases} \quad (3)$$

For the third component placement at the first column ( $i = 2, k = 0$ ):

$$\begin{cases} a_{20} = \frac{y_{\max} - d_y}{x_{\max} - x_0 - 2n - d_x} \\ a_{21} = \frac{y_{\max} - d_y}{x_{\max} - x_1 - 2n - d_x} \end{cases} \quad (4)$$

From (2)–(4), the general expression of the slope coefficient for the component coordinates is defined as follows:

$$a_{ij} = \frac{y_{\max} - d_y}{x_{\max} - x_j - i \times n - d_x} \quad (5)$$

where  $i = 0, 1, 2$ ;  $j = 0, 1, 2$ ; and  $n$  denotes the spacing between two consecutive components along the x-axis on a single PCB.

### 2.1.3. Extension to Multi-Column PCB Arrangement

Equation (5) only represents the slope coefficient  $a$  of the component coordinates for one column. However, since a printed circuit board (PCB) contains six different modules, we introduce  $k$  as the iteration index of components along the y-axis columns, as illustrated in Fig. 3. Accordingly, the slope coefficient  $a$  of the component coordinates on the PCB, at the first component iteration of the first column ( $i = 0, k = 0$ ), is determined as follows:

$$\begin{cases} a_{000} = \frac{y_{\max} - \eta y_{ijk} - d_y}{x_{\max} - x_0 - 0n - d_x} \\ a_{010} = \frac{y_{\max} - \eta y_{ijk} - d_y}{x_{\max} - x_1 - 0n - d_x} \end{cases} \quad (6)$$

Based on the rules derived for (5), the generalized form of (6) can be expressed as follows:

$$a_{ijk} = \frac{y_{\max} - \eta y_{ijk} - d_y}{x_{\max} - x_j - i \times n - d_x} \quad (7)$$

where  $\eta$  the amplification factor with respect to the repetition index  $k$  and the coordinates of the  $i$  –  $th$  column in the  $k$  –  $th$  sequence are defined in two cases as in Section 2.1.4.

#### 2.1.4. Even–Odd Indexing Rule for

For even values of  $k$  ( $2N$ ), the  $a_{ijk}$  coefficient of the component coordinate system is determined as follows:  $\eta = \frac{k}{2}$ , and the slope coefficient is:

$$a_{ijk(2N)} = \frac{y_{\max} - \eta y_{ijk} - d_y}{x_{\max} - x_j - i \times n - d_x} \quad (8)$$

Conversely, for odd values of  $k$  ( $2N + 1$ ), the  $a_{ijk}$  coefficient of the component coordinate system is determined as follows:  $\eta = \frac{k-1}{2}$ , and the slope coefficient becomes:

$$a_{ijk(2N+1)} = \frac{y_{\max} - \eta y_{ijk} - m - d_y}{x_{\max} - x_j - i \times n - d_x} \quad (9)$$

where  $N = 0, 1, 2, 3 \dots$  is an integer;  $k = 0, 1, 2, 3, 4, 5$  and  $m$  denotes the distance between two components along the  $y$ -axis between two printed circuit boards.

By substituting (8) and (9) into equation (1), the component coordinates are calculated as follows:

For even  $k = 2N$ :

$$y - (\eta y_{ijk} + d_y) = a_{ijk(2N)} [x - (x_j + i \times n + d_x)] \quad (10)$$

For odd  $k = 2N + 1$ :

$$y - (\eta y_{ijk} + m + d_y) = a_{ijk(2N+1)} [x - (x_j + i \times n + d_x)] \quad (11)$$

Equation (10) and (11) provide the final coordinate representation for component placement on the PCB, ensuring that the coordinate generation process remains scalable and systematically applicable across multiple PCB columns.

## 2.2. Component Coordinates of the IC:

After deriving the coordinate coefficients for component placement corresponding to  $k = 0, 1, 2, 3, 4, 5$ , this subsection extends the coordinate calculation framework to integrated circuits (ICs), which are arranged along the  $y$ -axis for  $k = 6, 7, 8$ . In this stage, the IC placement sequence follows the same transformation principle derived from (1), where the slope coefficient is recalculated to reflect the IC spacing and offset parameters.

### 2.2.1. Slope Coefficient Formulation for IC Placement

To determine the IC coordinate system, the slope coefficients are computed for each iteration index  $k$ . The derivations for  $k = 6, k = 7$ , and  $k = 8$  are presented below.

At the iteration corresponding to  $k = 6$ :

$$\begin{cases} a_{06} = \frac{y_{\max} - \frac{m}{2} - 0e - d_y}{x_{\max} - \alpha - n - 0f - d_x} \\ a_{16} = \frac{y_{\max} - \frac{m}{2} - 0e - d_y}{x_{\max} - \alpha - n - 1f - d_x} \end{cases} \quad (12)$$

At the iteration corresponding to  $k = 7$ :

$$\begin{cases} a_{07} = \frac{y_{\max} - \frac{m}{2} - 1e - d_y}{x_{\max} - \alpha - n - 0f - d_x} \\ a_{17} = \frac{y_{\max} - \frac{m}{2} - 1e - d_y}{x_{\max} - \alpha - n - 1f - d_x} \end{cases} \quad (13)$$

At the iteration corresponding to  $k = 8$ :

$$\begin{cases} a_{08} = \frac{y_{\max} - \frac{m}{2} - 2e - d_y}{x_{\max} - \alpha - n - 0f - d_x} \\ a_{18} = \frac{y_{\max} - \frac{m}{2} - 2e - d_y}{x_{\max} - \alpha - n - 1f - d_x} \end{cases} \quad (14)$$

### 2.2.2. Generalized Expression of the IC Slope Coefficient

From (12)–(14), it can be observed that the slope coefficient varies systematically with respect to the IC index  $k$ . Therefore, a generalized form of the slope coefficient for IC placement is established as:

$$a_{jk} = \frac{y_{\max} - \frac{m}{2} - \beta e - d_y}{x_{\max} - \alpha - n - j \times f - d_x} \quad (15)$$

where  $\beta$  is the amplification factor with respect to  $k$  for  $k > 5$ , therefore, the coefficient  $\beta$  is defined as follows:  $\beta = k - 6$ ;  $\alpha$  is an arbitrary positive coefficient of the IC along the x-axis;  $e$  is the spacing between two ICs along the y-axis; and  $f$  is the spacing between two ICs along the x-axis.

### 2.2.3. Coordinate System Representation of IC Placement

Using the same rules employed to derive (10) and (11) for (15), the coordinate system of the IC is defined as follows:

$$y - \left(\frac{m}{2} + \beta e + d_y\right) = a_{jk} [x - (\alpha + n + j \times f + d_x)] \quad (16)$$

Equation (16) provides a generalized linear representation of the IC placement coordinates, ensuring that the coordinate generation remains consistent and scalable for different IC positions along the placement sequence.

## 3. Feeder Coordinates and System Operational Cycle

Fig. 4 illustrates the schematic diagram of the coordinate system and the motion cycle of the axis. The diagram defines several key parameters governing the pick-and-place sequence. Here,  $l$  denotes the sequence of motion trajectories. The index  $\delta$  indicates the order of the feeders. The parameter  $d_f$  denotes the distance between consecutive feeders, and  $d_{f0}$  represents the distance from the origin to the first feeder ( $a_1$ ).

The motion cycle is executed according to the following rule: the trajectory where  $l = 0$  corresponds to the trajectory from the origin (O) to the position of feeder 5 supplying LED components ( $a_5$ ). Odd values of  $l$  represent the return cycles to the feeder for component pickup. Conversely, even values of  $l$  represent the forward cycles where the head moves toward the designated placement positions on the PCB.

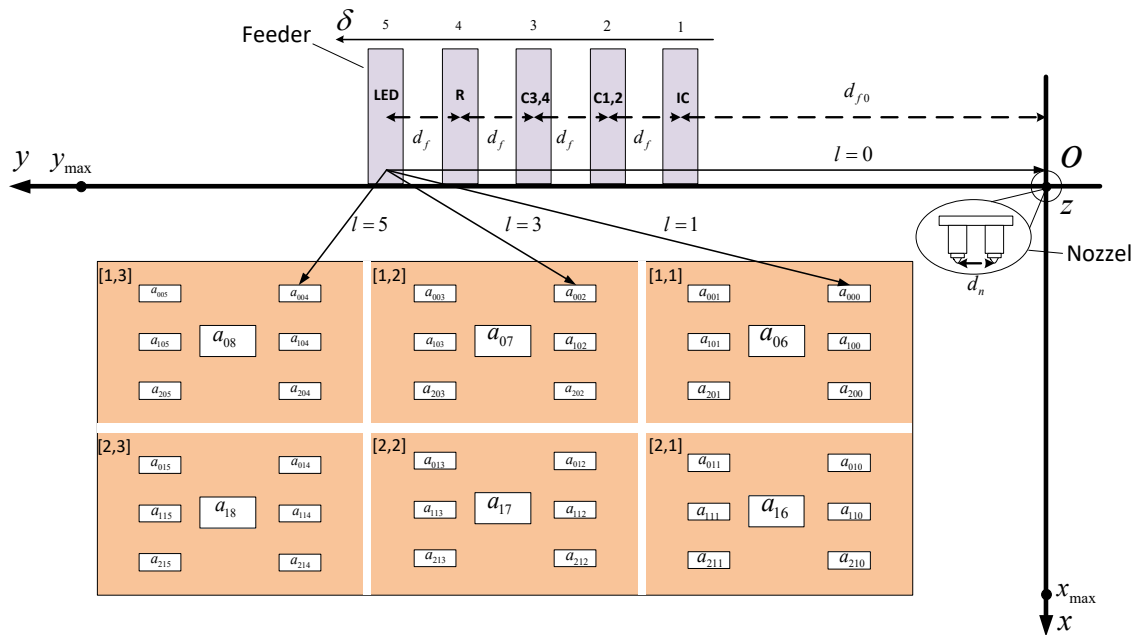


Fig. 4. Schematic diagram illustrating the feeder coordinates and the motion cycle of the axis

### 3.1. Feeder coordinates

In addition to the placement coordinates on the PCB, the pick-and-place process requires accurate pickup coordinates at the feeders. In the proposed framework, feeder pickup points are generated using the same linear relationship form employed for component coordinates. This ensures that feeder coordinate generation is consistent with the machine coordinate system and can be recalculated immediately when the datum point changes.

For the feeder supplying 5 (LED):

$$a_5 = \frac{y_{\max} - 4d_f - d_{f0}}{x_{\max}} \quad (17)$$

For the feeder supplying 4 (R):

$$a_4 = \frac{y_{\max} - 3d_f - d_{f0}}{x_{\max}} \quad (18)$$

For the feeder supplying 3 (C3 and C4):

$$a_3 = \frac{y_{\max} - 2d_f - d_{f0}}{x_{\max}} \quad (19)$$

For the feeder supplying 2 (C1 and C2):

$$a_2 = \frac{y_{\max} - d_f - d_{f0}}{x_{\max}} \quad (20)$$

For the feeder supplying 1 (IC):

$$a_1 = \frac{y_{\max} - 0d_f - d_{f0}}{x_{\max}} \quad (21)$$

Based on the rules for determining the coefficients in the generalized equations presented above, the generalized slope coefficient of the feeder coordinates is defined as follows:

$$a_{\delta} = \frac{y_{\max} - (\delta - 1)d_f - d_{f0}}{x_{\max}} \quad (22)$$

where  $\delta$  denotes the feeder index;  $d_f$  represents the distance between two feeders; and  $d_{f0}$  is the distance from the home position to the first feeder.

By substituting the slope coefficient in (22) into (1), the feeder coordinates are defined as follows:

$$y - ((\delta - 1)d_f + d_{f0}) = a_{\delta}x \quad (23)$$

### 3.2. Coordinates of the Primary and Secondary Nozzles with Respect to the Component and Feeder Coordinates

This section describes the dual-nozzle configuration utilized for the pick-and-place operation, with its schematic layout presented in Fig. 5. Within this configuration, a key geometric parameter is the fixed distance between the centers of the primary and secondary nozzles, denoted as  $d_n$ . This inter-nozzle distance is a critical design constant that directly influences the kinematic calculations, the planning of motion trajectories, and ultimately, the efficiency and accuracy of concurrent component handling.

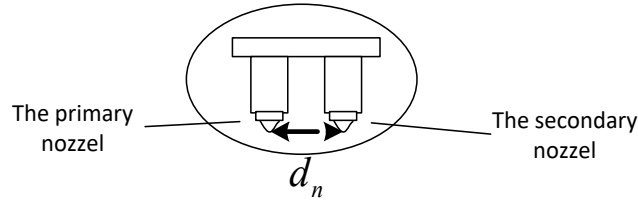


Fig. 5. Illustration of the dual nozzles and the distance between them

The primary nozzle is responsible for picking and placing components on the upper row of the printed circuit board (PCB), corresponding to  $j = 0$ . Meanwhile, the secondary nozzle handles the picking and placing of components on the lower row of the PCB, corresponding to  $j = 1$ . Accordingly, the equations describing the coefficients for all three coordinate systems (component, IC, and feeder) are reformulated as follows.

For the component coordinates, the coefficients of the primary and secondary nozzles are redefined as follows:

$$a_{ijk(2N)} = \frac{y_{\max} - \eta y_{ijk} - d_y - j d_n}{x_{\max} - x_j - i \times n - d_x} \quad (24)$$

$$a_{ijk(2N+1)} = \frac{y_{\max} - \eta y_{ijk} - m - d_y - j d_n}{x_{\max} - x_j - i \times n - d_x} \quad (25)$$

Thus, the component coordinates of the nozzle can be expressed as follows:

For even  $k = 2N$ :

$$y - (\eta y_{ijk} + d_y + j d_n) = a_{ijk(2N)}(x - (x_j + i \times n + d_x)) \quad (26)$$

For odd  $k = 2N + 1$ :

$$y - (\eta y_{ijk} + m + d_y + j d_n) = a_{ijk(2N+1)}(x - (x_j + i \times n + d_x)) \quad (27)$$

According to the IC coordinates, the slope coefficients of the primary and secondary nozzles are redefined as follows:

$$a_{jk} = \frac{y_{\max} - \frac{m}{2} - \beta e - d_y - jd_n}{x_{\max} - \alpha - n - j \times f - d_x} \quad (28)$$

Hence, the IC coordinates of the nozzle are given by:

$$y - \left(\frac{m}{2} + \beta e + d_y + jd_n\right) = a_{jk}(x - (\alpha + n + j \times f + d_x)) \quad (29)$$

With respect to the feeder coordinates, the slope coefficients of the nozzle are redefined as follows:

$$a_{j\delta} = \frac{y_{\max} - (\delta - 1)d_f - d_{f0} - jd_n}{x_{\max}} \quad (30)$$

Therefore, the feeder coordinates of the nozzle are expressed as:

$$y - ((\delta - 1)d_f + d_{f0} + jd_n) = a_{j\delta}x \quad (31)$$

#### 4. Operational Cycle of the Axis System

From the analysis of the two algorithm flowcharts in Fig. 6, the general equations for the slope coefficients of components, ICs, and feeders can be derived as follows:

Slope coefficient of the components:

$$a_{ijkl(2N)} = \frac{y_{\max} - \eta y_{ijk} - d_y - jd_n}{x_{\max} - x_j - i \times n - d_x} \quad (32)$$

$$a_{ijk(2N+1)} = \frac{y_{\max} - \eta y_{ijk} - m - d_y - jd_n}{x_{\max} - x_j - i \times n - d_x} \quad (33)$$

Slope coefficient of the IC:

$$a_{jkl} = \frac{y_{\max} - \frac{m}{2} - \beta e - d_y - jd_n}{x_{\max} - \alpha - n - j \times f - d_x} \quad (34)$$

Slope coefficient of the feeder:

$$a_{j\delta} = \frac{y_{\max} - (\delta - 1)d_f - d_{f0} - jd_n}{x_{\max}} \quad (35)$$

The component coordinates, derived from the slope coefficients in (32) and (33), are as follows:

$$\begin{cases} y - (\eta y_{ijk} + d_y + jd_n) = a_{ijkl(2N)}(x - (x_j + i \times n + d_x)) \\ y - (\eta y_{ijk} + m + d_y + jd_n) = a_{ijk(2N+1)}(x - (x_j + i \times n + d_x)) \end{cases} \quad (36)$$

The IC coordinates, derived from the slope coefficient in (34), are as follows:

$$y - \left(\frac{m}{2} + \beta e + d_y + jd_n\right) = a_{jkl}(x - (\alpha + n + j \times f + d_x)) \quad (37)$$

The feeder coordinates, derived from the slope coefficient in (35), are as follows:

$$y - ((\delta - 1)d_f + d_{f0} + jd_n) = a_{j\delta}x \quad (38)$$

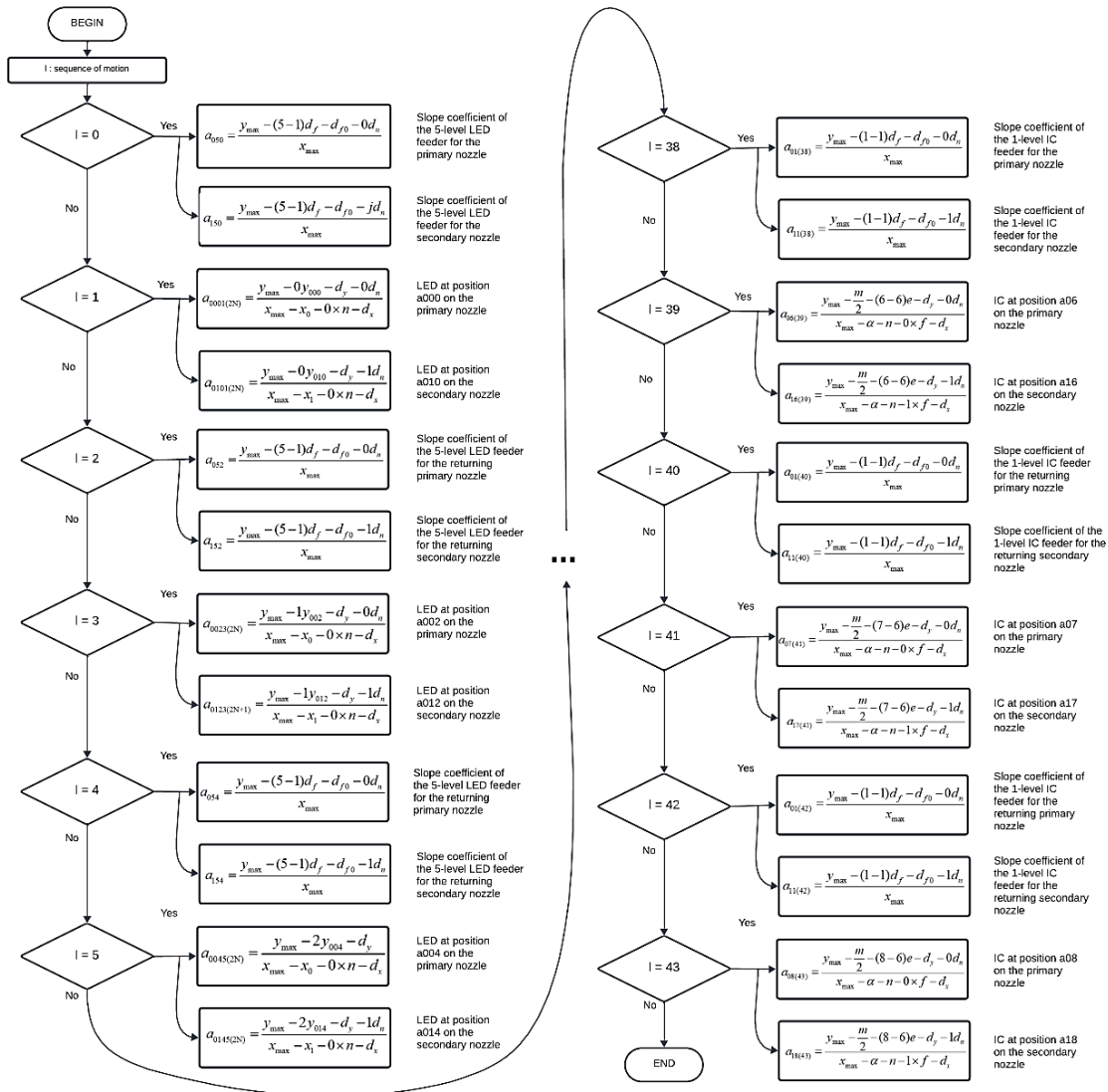


Fig. 6. Flowchart of the algorithm for the LED and IC operational cycle

Fig. 6 illustrates the overall operating cycle of the machine, in which each parameter  $a_{ijkl}$ , generalized in the previous section, is defined and applied specifically at each step of the process. The cycle is executed sequentially from  $l = 0$  to  $l = 43$ . At each step, the coordinates and slope coefficients of the components are determined, while the two nozzles (primary and secondary) are controlled in parallel to perform the pick-and-place operations. The computational equations indicate that the placement positions are derived from the feeder parameters, displacements, and design constants, thereby ensuring precision and stability during operation.

### 5. Results and Discussion

To validate the calculation results, experiments were conducted on a real PCB processing model, as illustrated in Fig. 7. The system comprises a Mitsubishi R04ENCPU PLC acting as the central controller, an MR-J4W3-222B amplifier for motion control, and three servo motors. Two 200W motors are employed for movement along the X and Y axes, while a 100W motor is responsible for movement along the Z-axis. Based on the coordinate values listed in Table 1 and Table 2, the position vectors are determined. To realize the motion toward these positions, the inverse kinematics problem of the robot is solved to derive the corresponding joint angles. High-speed positioning accuracy is

maintained through the implementation of a proportional-derivative (PD) controller for the servo motors. These components, together with the experimental model setup, are depicted in Fig. 7.

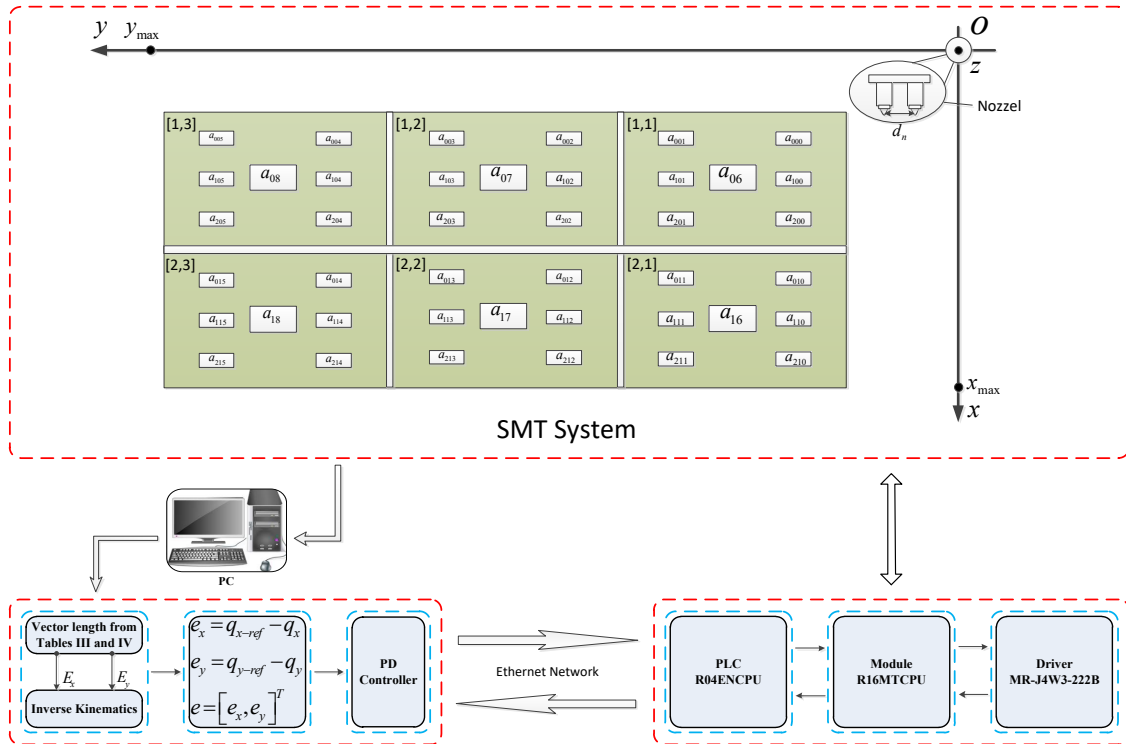


Fig. 7. Block diagram of device interconnections

From Table 1 and Table 2, the spatial length vectors corresponding to each target component were calculated. These geometric quantities were then converted into control signals for the SMT machine through inverse kinematics, ensuring that the pick-and-place system followed the correct trajectory and position. The experiment was conducted under laboratory conditions, with an average processing time of 3 minutes and 20 seconds per circuit board, as summarized in Table 3.

The authors designed a component assembly system based on SMT technology as illustrated in Fig. 2. The machine's technical parameters are defined in detail in Table 4. Based on that, all parameters were substituted into the general equations (32)–(38) to calculate and deduce the precise coordinates of the component, IC and feeder, the results of which are presented in Table 1 and Table 2.

Table 1. Component coordinates with respect to the machine coordinate system (mm)

Component Type	Table [1,1]		Table [1,2]		Table [1,3]		Table [2,1]		Table [2,2]		Table [2,3]	
	X	Y	X	Y	X	Y	X	Y	X	Y	X	Y
Led	179.7	33.6	179.7	66.6	179.7	99.61	199.9	80.6	199.9	113.6	199.9	146.6
	25	16	25	16	25	6	75	16	75	16	75	16
C3	184.5	33.6	184.5	66.6	184.5	99.61	204.7	80.6	204.7	113.6	204.7	146.6
	25	16	25	16	25	6	75	16	75	16	75	16
C4	189.3	33.6	189.3	66.6	189.3	99.61	209.5	80.6	209.5	113.6	209.5	146.6
	25	16	25	16	25	6	75	16	75	16	75	16
R	179.7	48.5	179.7	81.5	179.7	114.5	199.9	95.5	199.9	128.5	199.9	161.5
	25	2	25	2	25	2	75	2	75	2	75	2
C1	184.5	48.5	184.5	81.5	184.5	114.5	204.7	95.5	204.7	128.5	204.7	161.5
	25	2	25	2	25	2	75	2	75	2	75	2
C2	189.3	48.5	189.3	81.5	189.3	114.5	209.5	95.5	209.5	128.5	209.5	161.5
	25	2	25	2	25	2	75	2	75	2	75	2
IC	184.5	41.0	184.5	74.0	184.5	107.0	204.7	88.0	204.7	121.0	204.7	154.0
	25	68	25	68	25	68	75	68	75	68	75	68

**Table 2.** Coordinates of feeders relative to the machine coordinate system (mm)

Feeder Type	Slope $a$		Coordinates (X,Y)	
	The primary nozzles	The secondary nozzles	The primary nozzles	The secondary nozzles
Feeder 5 Led	0.555	0.4767	(0,267)	(0,314)
Feeder 4 (Capacitor C3, C4)	0.63	0.5517	(0,222)	(0,269)
Feeder 3 (Resistor R)	0.705	0.6267	(0,177)	(0,224)
Feeder 2 (Capacitor C1, C2)	0.78	0.7017	(0,132)	(0,179)
Feeder 1 (IC)	0.855	0.7767	(0,87)	(0,134)

**Table 3.** Detailed evaluation of errors and success rates of the model

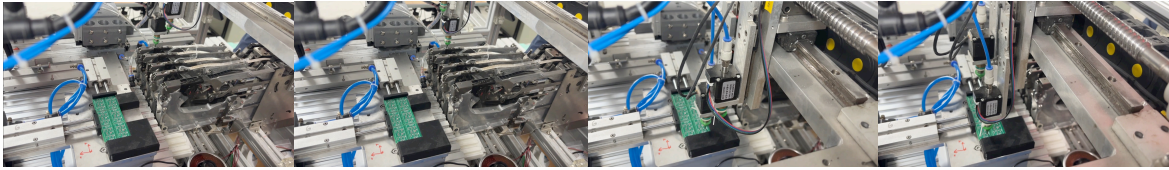
Trial No.	Number of Tested Circuits	Result	Notes
1	5	Completed	
2	10	Completed	
3	12	2 Faulty Circuits	The last two circuits missed components due to feeder malfunction.
4	20	3 Faulty Circuits	The first circuit missed a component due to pump malfunction. Circuits No. 11 and 12 lacked components due to feeder malfunction.
5	15	Completed	

**Table 4.** Parameters of the general equations (mm)

Variable	Value (mm)
$n$	4.8
$m$	14.904
$e$	33
$f$	20.25
$u$	20.25
$y_{ijk}$	33 (for all $ijk$ )
$x_{\max}$	600
$y_{\max}$	600
$d_f$	45
$d_{f0}$	87
$d_x$	179.725
$d_y$	33.616
$d_n$	47

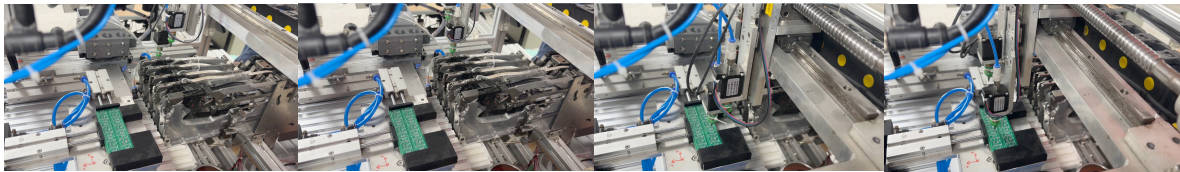
The computed results in Table 1 and Table 2 are embedded directly into the Mitsubishi R04ENCPU programmable logic controller (PLC) to verify and control the machining process in the physical testbed. These data serve as input to the control program, ensuring placement accuracy during assembly [53]–[56]. The system employs an MR-J4W3-222B driver to simultaneously control three servo motors, enabling high-precision positioning and speed regulation [57]–[60]. To clarify the integration between the computational module and the control hardware, the block diagram in Fig. 7 illustrates the linkage between the generated data and the PLC.

Fig. 8 is experiment on the XY-axis trajectory moving from the origin (0,0) to the feeder 5 supplying LED components and the placement points ( $a_{000} = (179.725, 33.616)$ ,  $a_{010} = (199.975, 80.616)$ ). The process begins at  $l = 0$  and ends at  $l = 1$ . During this sequence, the axis departs from the coordinate origin, where the primary nozzle first moves to Feeder 5 to pick up an LED, followed by the secondary nozzle performing the same pick-up operation. Once both nozzles have acquired components, the axis moves to the LED position  $a_{000}$ , where the primary nozzle places the component. Subsequently, the secondary nozzle proceeds to place its component.



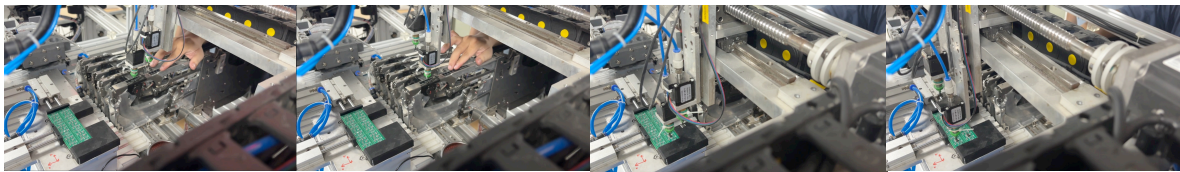
**Fig. 8.** Experimental illustration of the XY-axis trajectory from the origin (0,0) to Feeder 5 and the component placement positions, demonstrating the complete pick-and-place cycle for LED components

**Fig. 9** is experiment on the XY-axis trajectory from Feeder 5 supplying LED components to the designated placement points ( $a_{002} = (179.725, 66.616)$ ,  $a_{012} = (199.975, 113.616)$ ). The process runs from  $l = 0$  to  $l = 1$ , where the axis retrieves LED components from Feeder 5 and sequentially moves to place them at the specified positions using the main and sub nozzles.



**Fig. 9.** Experimental illustration of the XY-axis trajectory from Feeder 5 to the component placement positions, demonstrating the complete pick-and-place cycle for LED components

**Fig. 10** is experiment on the XY-axis trajectory from Feeder 1 supplying IC components to the designated placement points ( $a_{06} = (184.525, 41.068)$ ,  $a_{16} = (204.775, 88.068)$ ). The process runs from  $l = 42$  to  $l = 43$ , where the axis retrieves IC components from Feeder 1 and sequentially moves to place them at the specified positions using the main and sub nozzles.



**Fig. 10.** Experimental illustration of the XY-axis trajectory from Feeder 1 to the component placement positions, demonstrating the complete pick-and-place cycle for IC components

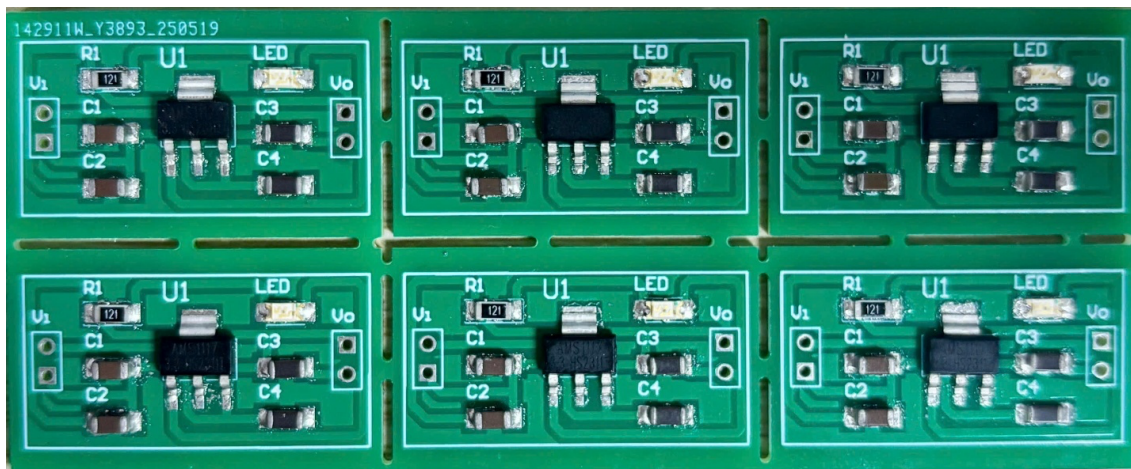
**Fig. 11** is final assembled PCB result with slope coefficient in [Table 5](#). The figure illustrates the completed product after the entire fabrication process. To evaluate the reliability of the proposed coordinate determination algorithm described in [Section 2](#), [Section 3](#), and [Section 4](#), five experimental runs were conducted, resulting in a total of 62 fabricated circuits. A summary of the experimental outcomes is provided in [Table 3](#).

Experimental results from [Table 3](#) indicate that the system achieved an accuracy of 91.94% (57/62 successful assemblies). However, in the unsuccessful trials, the primary causes were attributed to hardware malfunctions, while virtually no instances of misplacement or component deviation beyond the permissible tolerance (error  $\leq 0.2$  mm) were observed, as shown in [Fig. 11](#) in compliance with the IPC-A-610 standard. This confirms the reliability of the proposed coordinate determination algorithm presented in [Section 4](#). Compared with other approaches that typically rely on offline learning algorithms to optimize placement positions—yet lack adaptability to real-time positional variations. The method proposed in this study is developed and computed directly based on spatial coordinates and linear equations. This formulation enhances placement accuracy during operation. Future improvements could focus on enhancing hardware reliability, such as through more robust nozzle design or vision-based error correction. Furthermore, the underlying mathematical model shows potential for scalability, suggesting applicability to larger or more complex PCB layouts involving higher component density. The complete assembly process of a single circuit board is

illustrated in the accompanying demonstration video. (Link YouTube: <https://youtu.be/Lbx8E54XIk8>).

**Table 5.** Slope coefficients of components on the PCB

Component Type	Table [1,1]	Table [1,2]	Table [1,3]	Table [2,1]	Table [2,2]	Table [2,3]
	a	a	a	a	a	a
Led	1.3476	1.2691	1.1914	1.2984	1.2159	1.1334
C3	1.3632	1.2838	1.2046	1.3142	1.2307	1.1472
C4	1.3791	1.2988	1.2182	1.3303	1.2458	1.1613
R	1.3122	1.2336	1.1556	1.2611	1.1786	1.0961
C1	1.3273	1.2479	1.1684	1.2764	1.1929	1.1094
C2	1.3428	1.2625	1.1816	1.2922	1.2076	1.1231
IC	1.3466	1.2659	1.1864	1.2953	1.2118	1.1283



**Fig. 11.** The final step-down circuit after the fabrication process

## 6. Conclusion

This study proposed and validated a mathematical formulation for component coordinate computation in a Surface Mount Technology (SMT) placement system, aiming to enhance placement precision through deterministic coordinate transformation rather than purely heuristic trajectory tuning. Experimental evaluation on 62 printed circuit boards demonstrated that the proposed method achieved a placement accuracy of 91.94%, and importantly, no measured deviations exceeded the allowable tolerance threshold of 0.2 mm. These results confirm that the approach is not only theoretically consistent but also practically viable for improving placement reliability in real SMT production environments.

Despite these promising outcomes, several limitations should be recognized to ensure a balanced interpretation of the findings. First, the overall system performance remains partially constrained by hardware characteristics, particularly the use of dual nozzles and conventional feeders, which may introduce mechanical variability and limit repeatability under high-speed operation. Second, from a methodological standpoint, the proposed model has not yet been extensively verified across a wider range of PCB geometries, component densities, and layout complexities. Moreover, the current work does not include benchmarking against alternative coordinate computation or optimization approaches—such as heuristic-based strategies, Genetic Algorithms, or Ant Colony Optimization—which limits the ability to quantitatively position the method relative to established techniques in terms of cycle time reduction, setup time, and robustness.

Future research should therefore advance in two complementary directions. On the hardware side, system-level improvements such as upgrading the vacuum pump, implementing electrically driven

feeders, and improving nozzle precision are expected to reduce residual mechanical disturbances and enhance repeatability. On the methodological side, the proposed formulation should be extended and evaluated under more diverse PCB designs, scaled to larger production scenarios, and systematically compared against state-of-the-art algorithms and coordinate error compensation strategies. Addressing these aspects will not only improve accuracy and stability but also strengthen the industrial generalizability of the method and provide deeper insights into its applicability for next-generation SMT placement optimization.

**Author Contribution:** All authors contributed equally to the main contributor to this paper. All authors read and approved the final paper.

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